Perfect Conductor and Impedance Boundary Condition Corrections via a Finite Element Subproblem Method

Patrick Dular, Victor Péron, Ronan Perrussel, Laurent Krähenbühl, Christophe Geuzaine

1 University of Liège, Dept. of Electrical Engineering and Computer Science, ACE, B-4000 Liège, Belgium
2 F.R.S.-FNRS, Fonds de la Recherche Scientifique, Belgium
3 Université de Pau et des Pays de l’Adour, MAGIQUE3D (CNRS UMR5142-INRIA Bordeaux-Sud-Ouest), F-64013 Pau, France
4 Université de Toulouse, LAPLACE (CNRS UMR5213), F-31400 Toulouse, France
5 Université de Lyon, Ampère (CNRS UMR5005), École Centrale de Lyon, F-69134 Écully Cedex, France

Abstract—A finite element subproblem method is developed to correct the inaccuracies proper to perfect conductor and impedance boundary condition models, in particular near edges and corners of conductors, for a large range of conductivities and frequencies. Local corrections, supported by fine local meshes, can be obtained from each model to a more accurate one, allowing efficient extensions of their domains of validity.

I. INTRODUCTION

Magnetodynamic finite element (FE) modeling of conductors can be tackled at various levels of precision, e.g. considering them via perfect conductor or impedance boundary conditions (BCs). Avoiding to mesh their interior allows to lighten the computational efforts, which is interesting for the preliminary stage of a design. Perfect conductor BCs are suitable for high conductivities or frequencies, i.e. for low skin depths [1]. For larger skin depths, impedance BCs (IBCs) lead to a better accuracy. Such conditions are nevertheless generally based on analytical solutions of ideal problems and are therefore only valid in practice far from any geometrical discontinuities, e.g., edges and corners. Local modifications of IBCs can be defined (e.g. [2]).

It is here proposed to perform successive FE refinements via a subproblem (SP) method (SPM) [3] to correct the models with approximate BCs. Accurate skin and proximity effects, i.e. distributions of fields and current densities, are to be obtained for accurate force and Joule loss density distributions as well as for accurate interactions with neighboring regions. Sequences of SP corrections are developed for the magnetic vector potential FE magnetodynamic formulation. Each model can be included in the SP sequence and corrected by the other models of higher accuracy, with the advantage of using a different mesh at each step. The developed technique will be illustrated and validated on application examples.

II. COUPLED SUBPROBLEMS

A. Sequence of subproblems

To allow a natural progression from simple to more elaborate models, a complete problem is split into a series of SPs that define a sequence of changes, with the complete solution being replaced by the sum of the SP solutions [3]. Each SP is defined in its particular domain, generally distinct from the complete one and usually overlapping those of the other SPs. At the discrete level, this aims to decrease the problem complexity and to allow distinct meshes with suitable refinements and possible domain overlapping. The successive sources are obtained by means of Galerkin projections between the meshes. These have to be properly discretized to assure the conformity of all the sequenced FE weak formulations.

B. Canonical form of magnetodynamic subproblems

A canonical magnetodynamic problem, p, to be solved at step p of the SPM, is defined in a domain Ωp, with boundary ∂Ωp = Γp = Γh,fp ∪ Γs,fp. The eddy current conducting part of Ωp is denoted Ωc,fp and the non-conducting one Ωp \ Ωc,fp. Stranded conductors belong to Ωs⊂Ωc,fp. Magnetic field hp and electric current density jfp are related to magnetic flux density bfp and electric field efp, respectively, through the material relations

\[ h_{fp} = (\mu_p^{-1} - \mu_q^{-1}) b_{fp}, \quad j_{fp} = (\sigma_p - \sigma_q) e_{fp}, \]

(1a-b)

where \( \mu_p \) is the magnetic permeability, \( \sigma_p \) is the electric conductivity, and \( h_{fp} \) and \( j_{fp} \) are volume sources (VSs) defined by

\[ h_{fp} = (\mu_p^{-1} - \mu_q^{-1}) b_{fp}, \quad j_{fp} = (\sigma_p - \sigma_q) e_{fp}, \]

(2a-b)

for changes from \( \mu_q \) and \( \sigma_q \) for SP q to \( \mu_p \) and \( \sigma_p \) for SP p in some regions [3]. Also, BCs have to be defined for surface sources (SSs), possibly expressed from previous solutions, i.e.

\[ \mathbf{n} \times h_{fp} = f_{fp}, \quad \mathbf{n} \times b_{fp} = f_{fp}, \quad \mathbf{n} \times e_{fp} = f_{fp}, \]

(3a-b-c)

with \( \mathbf{n} \) the unit normal exterior to Ωp. Some paired portions of Γp can define double layers, with the thin layer in between exterior to Ωp; in particular, this will be the case with the perfect conductor and impedance BCs. They are denoted Γp+ and Γp− and are geometrically defined as a single surface γp with interface conditions (ICs), fixing the discontinuities \( (\cdot)_{fp} = (\cdot)_{fp}^+ - (\cdot)_{fp}^- \)

\[ \mathbf{n} \times h_{fp} = f_{fp}, \quad \mathbf{n} \times b_{fp} = f_{fp}, \quad \mathbf{n} \times e_{fp} = f_{fp}. \]

(4a-b-c)

With the magnetic vector potential \( \mathbf{a}_p \) and electric scalar potential \( \mathbf{v}_p \) defined via \( h_{fp} = \mathbf{curl} a_p \) and \( e_{fp} = -\mathbf{\nabla} v_p - \mathbf{grad} v_p \) and the resulting BC and IC

\[ \mathbf{n} \times \mathbf{a}_p = a_{fp}, \quad \mathbf{n} \times \mathbf{a}_p = a_{fp}, \]

(5a-b)

the \( \mathbf{v}_p \) weak formulation of the magnetodynamic problem is obtained from the weak form of the Ampère equation, i.e. [1]

\[
\begin{align*}
(u_{fp}^1 \mathbf{curl} a_p, \mathbf{curl} a_p)_{\Omega_p} + (h_{fp}, \mathbf{curl} a_p)_{\Omega_p} - (f_{fp}, a_p)_{\Omega_p} + & (\sigma_p \mathbf{\nabla} a_p, a_p)_{\Omega_p} + (\alpha_p \mathbf{u}_p, a_p)_{\Omega_p} < n \times h_{fp}, a_p >_{\Gamma_h,fp} \\
+ & < n \times \mathbf{h}_{fp}, a_p >_{\gamma_p} = 0, \forall a_p' \in L^1(\Omega_p)
\end{align*}
\]

(6)
where $F_p^1(\Omega_p)$ is a curl-conform function space defined on $\Omega_p$, gauged in $\Omega_{c,p}^{-1}$, and containing the basis functions for $a_p$ as well as for the test function $a'$ (at the discrete level, this space is defined by edge FEs; the gauge is based on the tree-co-tree technique); $(\cdot,\cdot)_{\Omega}$ and $<\cdot,\cdot>_{\Gamma}$ denote a volume integral in $\Omega$ and a surface integral on $\Gamma$, respectively, of the product of their field arguments.

III. CONDUCTOR MODELING – VARIOUS APPROXIMATIONS

A. Perfect conductor boundary condition (PCBC)

A SP ($p\equiv pc$) is defined in $\Omega_p$ by considering some conductors $\Omega_{c,p,i}$ ($i$ is the conductor index) as being perfect, i.e., of infinite conductivity ($\sigma_p\rightarrow \infty$) [1]. Its solution is thus independent of the conductivity and can serve as a reference solution for any conductivity further considered. This results in a zero skin depth and surface currents. The interior of $\Omega_{c,p,i}$ with zero fields inside, can thus be extracted from the studied domain $\Omega_p$ in (6) and treated via BCs (3b-c) fixing zero traces of $h_p$ and $e_p$ on their boundaries $\Gamma_{c,p,i}$, i.e.

$$\mathbf{n} \cdot \mathbf{h}_p|_{\Gamma_{c,p,i}} = 0, \quad \mathbf{n} \times \mathbf{e}_p|_{\Gamma_{c,p,i}} = 0 \text{ or } \mathbf{n} \times \mathbf{a}_p|_{\Gamma_{c,p,i}} = \mathbf{n} \times \text{grad} u_p|_{\Gamma_{c,p,i}} \quad (7a-b-c)$$

with $u_p$ any surface scalar potential. A non-zero trace $\mathbf{n} \times \mathbf{h}_p|_{\Gamma_{c,p,i}}$ will be part of the solution, thus giving a discontinuity $[\mathbf{n} \times \mathbf{h}_p]|_{\Gamma_{c,p,i}} = \mathbf{n} \times \mathbf{h}_p|_{\Gamma_{c,p,i}}$, i.e.

$$Z_{c,p,i} = (\sigma_p \mathbf{h}_p)^{-1} (1+j), \quad \delta_p = \sqrt{1/(\omega \sigma_p \mathbf{h}_p)} \quad (9a-b)$$

with $\omega$ the angular frequency ($\omega = 2\pi f$, with $f$ the frequency) and $j$ the imaginary unit ($\partial_\mathbf{n} = j \mathbf{e}_p$ in the frequency domain). BC (8) is thus to be expressed in (6) in term of the primal unknowns with $\mathbf{n} \times (\mathbf{n} \times \mathbf{e}_p)|_{\Gamma_{c,p,i}} = (\mathbf{n} \times (\partial_\mathbf{t} \mathbf{a}_p + u_p)) \times \mathbf{n}|_{\Gamma_{c,p,i}}$.

C. Modified impedance boundary condition (MIBC)

A modified IBC can be defined in the vicinity of corners in 2-D or edges in 3-D, as given by a reference problem [2]. If $\Gamma_{c,p,i}$ has one corner singularity located at the origin $X=0$, of angle $\beta$ in $\Omega_{c,p,i}$, then the scaling $X=x/\delta_p$ gives a “profile” term $V_\beta$, that is independent of $\delta_p$ and satisfies a reference problem described in details in [2], with a reference corner of the same opening $\beta$ as in $\Omega_{c,p,i}$. The surface impedance close to the corner can be then approximated by

$$Z_{c,p,i} \approx Z_{c,p,i} \left(\frac{\delta_p}{\delta_p}ight)^{1/2} \left(\frac{\delta_p}{\delta_p}\right)^{1/2} V_\beta \left(\frac{\delta_p}{\delta_p}\right)^{1/2} \frac{\delta_p}{\delta_p} \cdot (10)$$

D. Corrections of BCs up to volume conductors

Any SP $p$ can be defined as a correction of a previous SP $q$, without involving the already considered sources (e.g. inductors, previous VSs and SSs). For a change from a BC approximation to another, SSs have to be defined. E.g., from a PCBC-SP $q$ to an IBC-SP $p$, SS $\mathbf{n} \times \mathbf{h}_q|_{\Gamma_{c,p,i}}$ is to be subtracted from the right hand side of (8) for SP $p$. It is thus involved in a surface integral term in (6), to be weakly expressed from (6) for SP $q$, i.e. generally limited to

$$<\mathbf{n} \times \mathbf{h}_q, a'>_{\Gamma_{c,p,i}} \equiv -\left(\mathbf{u}_q^\dagger \text{curl} \mathbf{a}_q, \text{curl} a'\right)_{\Omega_q} \quad (11)$$

A change from a PCBC-SP to an accurate volume model has been developed in [1] and is here included in an extended correction procedure, with intermediary IBC-SP or MIBC-SP corrections. A change from an IBC-SP $q$ to a volume SP $p$ can involve both SSs and VSs. Considering a zero solution in $\Omega_{c,p,i}$, and thus carrying all the fields in the double layer of its boundary, trace discontinuities of both $\mathbf{h}_q$ and $\mathbf{e}_q$ ($a_q$) occur: their opposite values then define SSs for SP $p$ in (4a), weakly expressed as (11), and (4c) (and (5b)), strongly expressed in function space $F_p^1(\Omega)$. Also, a zero solution $q$ in $\Omega_{c,p,i}$ allows the VSs (2-a-b) to be zero. VSs have thus to be only considered for further local changes of $\mathbf{h}_q$ and $\mathbf{e}_p$, e.g. for nonlinear magnetic materials or temperature dependent conductivities.

Fig. 1 shows some correction solutions for PCBC, IBC and MIBC models, with increasing accuracy. Fig. 2 shows the corresponding surface impedances before correction.

IV. REFERENCES